

Transmission Lines

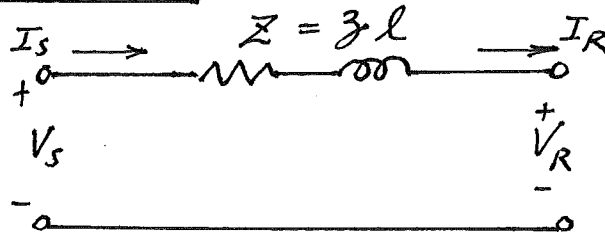
Equivalent Circuits:

The short line approximation: for less than 80 km (50 miles)

The Medium Line approximation: for 80 ~ 250 km (150 miles)

The Long Line approximation: for longer than 250 km

1. Short Line: < 80 km (50 miles)



$$Z = R + j\omega L \quad \Omega/\text{m}$$

$$Z = Zl$$

l : length in [m]

V_s : line-to-neutral voltage.

Relation between the sending-end and the receiving-end variables:

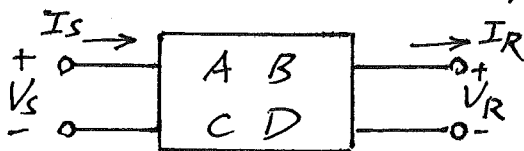
$$V_s = V_R + Z I_R$$

$$I_s = I_R$$

In a matrix form,

$$\begin{bmatrix} V_s \\ I_s \end{bmatrix} = \begin{bmatrix} 1 & Z \\ 0 & 1 \end{bmatrix} \begin{bmatrix} V_R \\ I_R \end{bmatrix} = \begin{bmatrix} A & B \\ C & D \end{bmatrix} \begin{bmatrix} V_R \\ I_R \end{bmatrix}$$

Generalized two-port network:



$$V_s = A V_R + B I_R$$

$$I_s = C V_R + D I_R$$

NOTE: For linear, passive, bilateral network,

$$AD - BC = 1$$

For symmetric network,

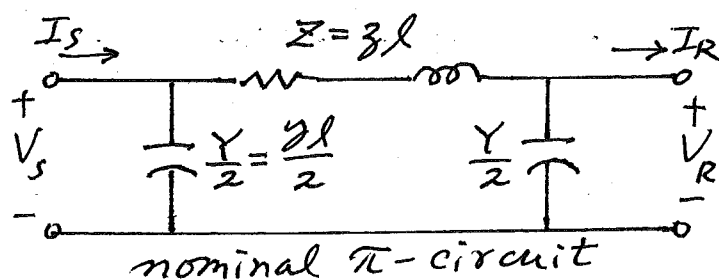
$$A = D$$

Inverse relationship:

Since the determinant of the (A, B, C, D) matrix is $AD - BC = 1$, we have

$$\begin{bmatrix} V_R \\ I_R \end{bmatrix} = \begin{bmatrix} D & -B \\ -C & A \end{bmatrix} \begin{bmatrix} V_S \\ I_S \end{bmatrix}$$

2. Medium-length Line: 80 ~ 250 km (50 ~ 150 miles)



Shunt admittance:

$$y = G + j\omega C \text{ S/m}$$

$$Y = yL$$

$$V_S = V_R + Z \left(I_R + \frac{Y}{2} V_R \right) = \left(1 + \frac{ZY}{2} \right) V_R + Z I_R$$

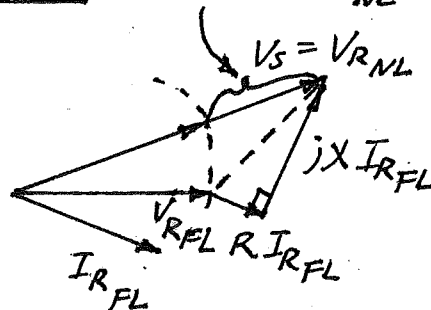
$$\begin{aligned} I_S &= I_R + \frac{Y}{2} V_R + \frac{Y}{2} V_S \\ &= I_R + \frac{Y}{2} V_R + \frac{Y}{2} \left(1 + \frac{ZY}{2} \right) V_R + \frac{Y}{2} Z I_R \\ &= Y \left(1 + \frac{ZY}{4} \right) V_R + \left(1 + \frac{ZY}{2} \right) I_R \end{aligned}$$

$$\therefore \begin{bmatrix} V_S \\ I_S \end{bmatrix} = \begin{bmatrix} 1 + \frac{ZY}{2} & Z \\ Y \left(1 + \frac{ZY}{4} \right) & 1 + \frac{ZY}{2} \end{bmatrix} \begin{bmatrix} V_R \\ I_R \end{bmatrix}$$

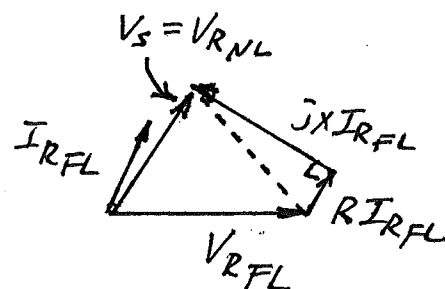
3. Voltage Regulation:

$$\% VR = \frac{|V_{RNL}| - |V_{RFL}|}{|V_{RFL}|} \times 100$$

Short Line: $\Delta V = |V_{RNL}| - |V_{RFL}|$



(a) lagging p.f.
% VR is positive



(b) leading p.f.
% VR is negative

Medium Line:

$$V_s = A V_R + B I_R, \quad I_R = 0 \text{ for No Load}$$

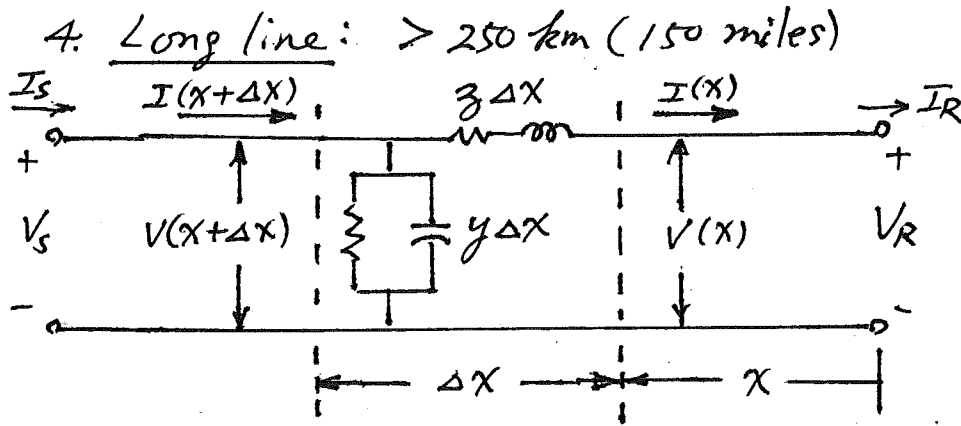
$$\Rightarrow V_s = A V_{RNL}$$

$$\text{or } V_{RNL} = \frac{V_s}{A} \quad (= V_s \text{ for short line})$$

For EHV lines $\pm 5\%$ of rated voltage, i.e.,
10% VR is acceptable.

Line loadability is determined by

- Thermal limit: for short line
- Voltage-drop limit: for longer line upto 300 km
- Steady-state stability limit: over 300 km



z : series impedance Ω/m

y : shunt admittance S/m

For a small section of length Δx ,

$$V(x+\Delta x) = V(x) + (z\Delta x) I(x)$$

$$I(x+\Delta x) = I(x) + (y\Delta x) V(x)$$

$$\Rightarrow \Delta V = V(x+\Delta x) - V(x) = (z\Delta x) I(x)$$

$$\Delta I = I(x+\Delta x) - I(x) = (y\Delta x) V(x)$$

As $\Delta x \rightarrow dx$,

$$dV = I z dx$$

$$dI = V y dx$$

$$\text{or } \frac{dV}{dx} = z I(x)$$

$$\frac{dI}{dx} = y V(x) \quad (1)$$

Differentiating with respect to x ,

$$\frac{d^2 V}{dx^2} = z \frac{dI}{dx}$$

$$\frac{d^2 I}{dx^2} = y \frac{dV}{dx}$$

$$\text{or } \frac{d^2 V}{dx^2} = zy V(x)$$

$$\frac{d^2 I}{dx^2} = zy I(x)$$

: These are wave equations

Solution is in the form of

$$V(x) = A e^{\sqrt{zy} x} + B e^{-\sqrt{zy} x}$$

Note: $\frac{dV}{dx} = \sqrt{3Y} (A e^{\sqrt{3Y}x} - B e^{-\sqrt{3Y}x})$ (2)

$$\frac{d^2V}{dx^2} = 3Y (A e^{\sqrt{3Y}x} + B e^{-\sqrt{3Y}x})$$

$= 3Y V(x)$, thus satisfying the wave equation.

Since $V(x)$ and $I(x)$ are coupled by (1), with (2),

$$I(x) = \frac{1}{3} \frac{dV}{dx}$$

$$= \frac{1}{\sqrt{\frac{3}{Y}}} (A e^{\sqrt{3Y}x} - B e^{-\sqrt{3Y}x})$$

Define line parameters:

$$\gamma = \sqrt{3Y} \text{ [m}^{-1}\text{]}, \text{ propagation constant}$$

$$= \alpha + j\beta, \quad \alpha = \text{attenuation constant}$$

$$\beta = \text{phase constant}$$

$$Z_c = \sqrt{\frac{3}{Y}} \text{ } [\Omega], \text{ characteristic impedance}$$

Then, the voltage and current at x are:

$$V(x) = A e^{\gamma x} + B e^{-\gamma x} \quad (3)$$

$$I(x) = \frac{1}{Z_c} (A e^{\gamma x} - B e^{-\gamma x})$$

To find constants A and B , apply boundary conditions at $x=0$:

$$V_R = V(0) = A + B$$

$$I_R = \frac{1}{Z_c} (A - B)$$

Solving for A & B,

$$A = \frac{V_R + Z_c I_R}{2}$$

$$B = \frac{V_R - Z_c I_R}{2}$$

Substituting these in (3),

$$V(x) = \frac{V_R + Z_c I_R}{2} e^{\gamma x} + \frac{V_R - Z_c I_R}{2} e^{-\gamma x} \quad (4)$$

$$I(x) = \underbrace{\frac{V_R/Z_c + I_R}{2} e^{\gamma x}}_{\text{incident wave}} - \underbrace{\frac{V_R/Z_c - I_R}{2} e^{-\gamma x}}_{\text{reflected wave}}$$

NOTE: If $I_R = 0$ (open circuit, no load)

$$V(0)_{\text{incident}} = V(0)_{\text{reflected}} = \frac{V_R}{2}$$

$$I(0)_{\text{incident}} = -I(0)_{\text{reflected}} = -\frac{V_R}{2Z_c}$$

$$\Rightarrow V(0) = \frac{V_R}{2} + \frac{V_R}{2} = V_R$$

$$I(0) = \frac{V_R}{2Z_c} - \frac{V_R}{2Z_c} = 0$$

NOTE: If the line is terminated with characteristic impedance, Z_c , then

$$V_R = Z_c I_R$$

⇒ There is no reflected waves, which is called "Impedance Matching."

Impedance matching is common in communication systems. In power systems

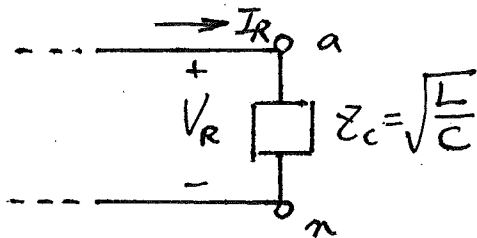
$$Z_c = \sim 400 \Omega \angle \sim 15^\circ \text{ for single circuit}$$

$$\sim 200 \Omega \angle \sim 15^\circ \text{ for parallel circuit}$$

For lossless line (series resistance and shunt conductance are zero),

$$Z_c = \sqrt{\frac{Z}{Y}} = \sqrt{\frac{j\omega L}{j\omega C}} = \sqrt{\frac{L}{C}}, \text{ surge impedance}$$

Surge Impedance Loading (SIL)



$$V_R = \frac{V_L}{\sqrt{3}}$$

$$I_R = \frac{V_R}{Z_c} = \frac{V_L}{\sqrt{3} Z_c}$$

$$SIL = \sqrt{3} V_L I_R = \frac{V_L^2}{Z_c}$$

$$= \frac{V_{\text{rated, line-to-line}}^2}{Z_c}$$

Wave Propagation:

$V(x)$ and $I(x)$ propagate (4) with the propagation constant

$$\gamma = \sqrt{ZY} = \alpha + j\beta$$

For loss-less line, $\alpha = 0$ and

$$\gamma = \sqrt{(j\omega L)(j\omega C)} = j\omega\sqrt{LC} = j\beta \text{ [m}^{-1}\text{]}$$

β : phase shift in rad/m.

A wave length is the distance required to change the phase of $V(x)$ or $I(x)$ by 2π radians or 360° , i.e.,

$$\lambda\beta = 2\pi,$$

$$\Rightarrow \text{Wave length, } \lambda = \frac{2\pi}{\beta} = \frac{2\pi}{\omega\sqrt{LC}} = \frac{1}{f\sqrt{LC}} \text{ [m]}$$

$$= \frac{3 \times 10^8}{60} = 5 \times 10^6 \text{ m} = 5000 \text{ km}$$

$$= 3100 \text{ mi}$$

Velocity of propagation:

$$v = f \cdot \lambda = \frac{1}{\sqrt{LC}} \text{ [m/s]}$$

$$= \frac{1}{\sqrt{\frac{\mu_0}{2\pi} \ln \frac{D}{r} \cdot \frac{2\pi\epsilon_0}{\ln \frac{D}{r}}}} = \frac{1}{\sqrt{\mu_0 \epsilon_0}} \approx 3 \times 10^8 \text{ m/s}$$

: speed of light in free space.

For cables, $\frac{\epsilon}{\epsilon_0} \approx 3 \sim 5$, implying that

v is lower than that for overhead lines.

Long Line Model in Hyperbolic Form:

Recall that

$$\sinh \theta = \frac{e^{\theta} - e^{-\theta}}{2}, \quad \cosh \theta = \frac{e^{\theta} + e^{-\theta}}{2}$$

$$\left[\sin \theta = \frac{e^{j\theta} - e^{-j\theta}}{2j}, \quad \cos \theta = \frac{e^{j\theta} + e^{-j\theta}}{2} \right]$$

$$\Rightarrow \sinh \theta + \cosh \theta = e^{\theta}, \quad \cosh \theta - \sinh \theta = e^{-\theta}$$

Thus, from (4),

$$\begin{aligned} V(x) &= \frac{V_R + Z_c I_R}{2} e^{\gamma x} + \frac{V_R - Z_c I_R}{2} e^{-\gamma x} \\ &= \frac{V_R + Z_c I_R}{2} (\sinh \gamma x + \cosh \gamma x) + \frac{V_R - Z_c I_R}{2} (\cosh \gamma x - \sinh \gamma x) \end{aligned}$$

$$= \cosh(\gamma x) V_R + Z_c \sinh(\gamma x) I_R$$

$$\begin{aligned} I(x) &= \frac{V_R/Z_c + I_R}{2} e^{\gamma x} - \frac{V_R/Z_c - I_R}{2} e^{-\gamma x} \\ &= \frac{V_R/Z_c + I_R}{2} (\sinh \gamma x + \cosh \gamma x) - \frac{V_R/Z_c - I_R}{2} (\cosh \gamma x - \sinh \gamma x) \\ &= \frac{\sinh(\gamma x)}{Z_c} V_R + \cosh(\gamma x) I_R \end{aligned}$$

$$\therefore \begin{bmatrix} V(x) \\ I(x) \end{bmatrix} = \begin{bmatrix} A(x) & B(x) \\ C(x) & D(x) \end{bmatrix} \begin{bmatrix} V_R \\ I_R \end{bmatrix}$$

where

$$A(x) = D(x) = \cosh(\gamma x) \quad [\text{pu}]$$

$$B(x) = Z_c \sinh(\gamma x) \quad [\Omega]$$

$$C(x) = \frac{1}{Z_c} \sinh(\gamma x) \quad [\text{S}]$$

Finally, at $x=l$, $V(l)=V_s$, $I(l)=I_s$, and

$$A = D = \cosh(\gamma l) \quad [\text{pu}]$$

$$B = Z_c \sinh(\gamma l) \quad [\Omega]$$

$$C = \frac{1}{Z_c} \sinh(\gamma l) \quad [\text{S}]$$

Computation of Hyperbolic Functions:

Recall that

$$\gamma = \sqrt{zy} = \alpha + j\beta \quad [\text{m}^{-1}], \text{ complex}$$

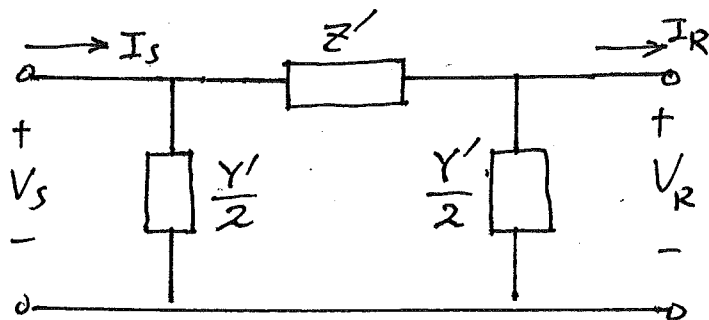
$$\text{Thus, } e^{\gamma l} = e^{(\alpha + j\beta)l} = e^{\alpha l} e^{j\beta l} = e^{\alpha l} \angle \beta l$$

Then,

$$\cosh(\gamma l) = \frac{e^{\gamma l} + e^{-\gamma l}}{2} = \frac{1}{2} (e^{\alpha l} \angle \beta l + e^{-\alpha l} \angle -\beta l)$$

$$\sinh(\gamma l) = \frac{e^{\gamma l} - e^{-\gamma l}}{2} = \frac{1}{2} (e^{\alpha l} \angle \beta l - e^{-\alpha l} \angle -\beta l)$$

Equivalent π -Circuit:



Recall in the nominal π -equivalent circuit

$$A = D = 1 + \frac{Y'Z'}{2}$$

$$B = Z'$$

$$C = Y' \left(1 + \frac{Y'Z'}{4} \right)$$

Equating these with (A, B, C, D)

parameters in the long-line representation,

$$A: 1 + \frac{Y'Z'}{2} = \cosh \gamma l$$

$$B: Z' = Z_c \sinh \gamma l$$

$$C: Y' \left(1 + \frac{Y'Z'}{4} \right) = \frac{\sinh \gamma l}{Z_c}$$

Solving B for Z' ,

$$Z' = \sqrt{\frac{Z}{Y}} \sinh \gamma l = Z l \frac{\sinh \gamma l}{\sqrt{ZY} l} = Z \underbrace{\frac{\sinh \gamma l}{\gamma l}}_{F_1} = Z F_1$$

where $Z = Z l$, total series impedance, F_1 , correction factor

Solving A for Y' ,

$$\begin{aligned} \frac{Y'}{2} &= \frac{1}{Z_c} \frac{\cosh \gamma l - 1}{\sinh \gamma l} = \sqrt{\frac{Y}{Z}} \tanh \frac{\gamma l}{2} \\ &= \frac{\gamma l}{2} \frac{\tanh \frac{\gamma l}{2}}{\sqrt{ZY} l/2} = \frac{Y}{2} \underbrace{\frac{\tanh \frac{\gamma l}{2}}{\gamma l/2}}_{F_2} \\ &= \frac{Y}{2} F_2 \end{aligned}$$

where $Y = \gamma l$, total shunt admittance.

NOTE: For small γl , $Z' = Z$, $Y' = Y$.