ELC 5396: Digital Communications

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2 Fast Fourier Transform Algorithms

$$X(\omega) = \sum_{n=-\infty}^{\infty} x(n) e^{-j\omega n}, \quad \omega \in [-\pi,\pi) \text{ or } \omega \in [0,2\pi)$$

Synthesis Equation

$$X(n) = rac{1}{2\pi} \int_{-\pi}^{\pi} X(\omega) e^{j\omega n} d\omega$$

 $X(\omega)$ is periodic with period 2π .

The Discrete Fourier Transform (DFT)

N-point DFT.

Analysis Equation

$$X(k) = \sum_{n=0}^{N-1} x(n) e^{-j2\pi \frac{k}{N}n}, \qquad k = 0, 1, 2, \dots, N-1$$

Synthesis Equation

$$x(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) e^{j2\pi \frac{k}{N}n}, \quad n = 0, 1, 2, \dots, N-1$$

N samples of the Fourier transform at N equally spaced frequencies. $\omega_k = \frac{2\pi k}{N}$, k = 0, 1, 2, ..., N - 1.

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$$X(k) = \sum_{n=0}^{N-1} x(n) W_N^{kn}, \qquad k = 0, 1, 2, \dots, N-1$$

Synthesis Equation

$$x(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) W_N^{-kn}, \qquad n = 0, 1, 2, \dots, N-1$$

where, $W_N = e^{-j2\pi/N}$.

$$X(k) = \sum_{n=0}^{N-1} x(n) W_N^{kn}, \qquad k = 0, 1, 2, \dots, N-1$$

To calculate one frequency sample (each k) in the analysis equation (direct Fourier transform), we need N complex multiplications and N-1 complex additions.

For all N frequency samples, we need a total of N^2 complex multiplications and N(N-1) complex additions.

$$X(k) = \sum_{n=0}^{N-1} x(n) W_N^{kn}, \qquad k = 0, 1, 2, \dots, N-1$$

For all N frequency samples, we need a total of $4N^2$ real multiplications and N(4N-2) real additions.

$$X(k) = \sum_{n=0}^{N-1} x(n) W_N^{kn}, \qquad k = 0, 1, 2, \dots, N-1$$

The computation complexity of the DFT is proportional to N^2 .

As a comparison, the computational complexity of the FFT is proportional to $N \log N$.

Use the periodicity of the sequence W_N^{kn} to reduce computation.

$$\mathcal{W}^{kN}_{\mathcal{N}}=e^{-jrac{2\pi}{\mathcal{N}}k\mathcal{N}}=e^{-j2\pi k}=1$$

$$W_N^{k(N-n)} = W_N^{-kn} = (W_N^{kn})^*$$

(complex conjugate symmetry)

$$W_N^{kn} = W_N^{k(n+N)} = W_N^{(k+N)n}$$

(periodicity)

Considering N an integer power of 2, i.e. $N = 2^{\nu}$.

$$X(k) = \sum_{n=0}^{N-1} x(n) W_N^{nk}, \quad k = 0, 1, \dots, N-1$$

=
$$\sum_{n \text{ even}} x(n) W_N^{nk} + \sum_{n \text{ odd}} x(n) W_N^{nk}$$

=
$$\sum_{r=0}^{N/2-1} x(2r) W_N^{2rk} + \sum_{r=0}^{N/2-1} x(2r+1) W_N^{(2r+1)k}$$

=
$$\sum_{r=0}^{N/2-1} x(2r) (W_N^2)^{rk} + W_N^k \sum_{r=0}^{N/2-1} x(2r+1) (W_N^2)^{rk}$$

$$W_N^2 = e^{-2j\frac{2\pi}{N}} = e^{-j\frac{2\pi}{N/2}} = W_{N/2}$$

Therefore,

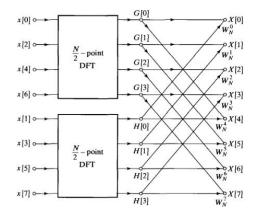
$$X(k) = \sum_{r=0}^{N/2-1} x(2r)(W_N^2)^{rk} + W_N^k \sum_{r=0}^{N/2-1} x(2r+1)(W_N^2)^{rk}$$

=
$$\sum_{r=0}^{N/2-1} x(2r)W_{N/2}^{rk} + W_N^k \sum_{r=0}^{N/2-1} x(2r+1)W_{N/2}^{rk}$$

=
$$G(k) + W_N^k H(k)$$

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G(k) is an (N/2)-point DFT of even samples x(2r); H(k) is an (N/2)-point DFT of odd samples x(2r + 1).



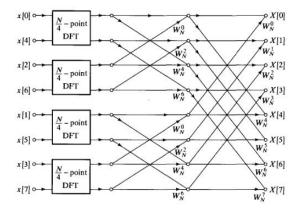
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$$G(k) = \sum_{l=0}^{N/4-1} g(2l) W_{N/4}^{lk} + W_{N/2}^{k} \sum_{l=0}^{N/4-1} g(2l+1) W_{N/4}^{lk}$$

$$H(k) = \underbrace{\sum_{l=0}^{N/4-1} h(2l) W_{N/4}^{lk}}_{(N/4)-\text{point DFT}} + W_{N/2}^{k} \underbrace{\sum_{l=0}^{N/4-1} h(2l+1) W_{N/4}^{lk}}_{(N/4)-\text{point DFT}}$$

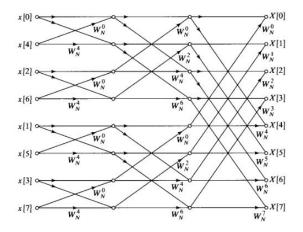
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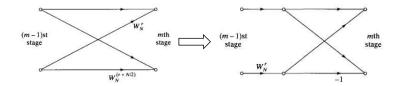
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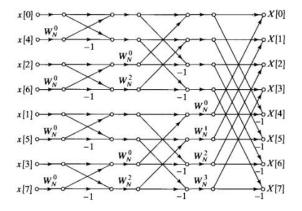
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$$W_N^{N/2} = e^{-j\frac{2\pi}{N}\frac{N}{2}} = e^{-j\pi} = -1$$
$$W_N^{r+N/2} = W_N^{N/2}W_N^r = -W_N^r$$



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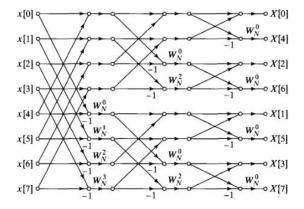
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x(n)'s index n	binary
0	000
4	100
2	010
6	110
1	001
5	101
3	011
7	111

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Decimation-in-Frequency Fast Fourier Transform (FFT)



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