## ELC 4351: Digital Signal Processing

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Freq Analysis of Signals II

## Frequency Analysis of Signals

Frequency-Domain and Time-Domain Signal Properties Frequency-Domain and Time-Domain Signal Properties

Properties of the Fourier Transform for Discrete-Time Signals Symmetry Properties of the Fourier Transform Fourier Transform Theorems and Properties

## Frequency-Domain and Time-Domain Signal Properties

Frequency Analysis Tools	
The Fourier series	for continuous-time periodic signals
The Fourier transform	for continuous-time aperiodic signals
The Fourier series	for discrete-time periodic signals
The Fourier transform	for discrete-time aperiodic signals

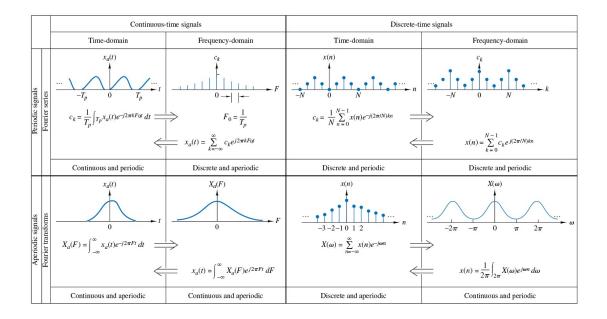
Continuous-time signals have aperiodic spectra

Discrete-time signals have periodic spectra

Periodic signals have discrete spectra

Aperiodic finite energy signals have continuous spectra

### The Fourier Series for Continuous-Time Periodic Signals



Periodicity with period  $\alpha$  in one domain implies discretization with spacing  $1/\alpha$  in the other domain, and *vice versa*.

$$\begin{split} X(\omega) &\triangleq F\{x(n)\} = \sum_{n=-\infty}^{\infty} x(n)e^{-j\omega n} \\ x(n) &\triangleq F^{-1}\{X(\omega)\} = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(\omega)e^{j\omega n} d\omega \end{split}$$

Fourier transform pair:  $x(n) \longleftrightarrow^F X(\omega)$ 

where,  $X(\omega)$  is periodic with period  $2\pi$ .

If signal is complex, it can be expressed in rectangular form

$$\begin{aligned} x(n) &= x_R(n) + jx_I(n) \\ X(\omega) &= X_R(\omega) + jX_I(\omega) \end{aligned}$$

### Symmetry Properties of the Fourier Transform

When a signal satisfies some symmetry properties in the time domain, these properties impose some symmetry conditions on its Fourier transform.

Using the rectangular form and  $e^{j\omega} = \cos \omega + j \sin \omega$ , we have

$$X_R(\omega) = \sum_{n=-\infty}^{\infty} [x_R(n)\cos\omega n + x_I(n)\sin\omega n]$$
  
$$X_I(\omega) = -\sum_{n=-\infty}^{\infty} [x_R(n)\sin\omega n - x_I(n)\cos\omega n]$$

and

$$x_R(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} [X_R(\omega) \cos \omega n - X_I(\omega) \sin \omega n] d\omega$$
  
$$x_I(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} [X_R(\omega) \sin \omega n + X_I(\omega) \cos \omega n] d\omega$$

Real signals.  $x_R(n) = x(n)$  and  $x_I(n) = 0$ .

$$X_R(\omega) = \sum_{n=-\infty}^{\infty} x(n) \cos \omega n$$
$$X_I(\omega) = -\sum_{n=-\infty}^{\infty} x(n) \sin \omega n$$

It follows that

$$X_R(-\omega) = X_R(\omega)$$
  
$$X_I(-\omega) = -X_I(\omega)$$

 $\implies X^*(\omega) = X(-\omega)$ . The spectrum of a real signal has *Hermitian symmetry*.

### Symmetry Properties of the Fourier Transform

Real signals.  $x_R(n) = x(n)$  and  $x_I(n) = 0$ .

$$X_R(-\omega) = X_R(\omega)$$
 (even)  
 $X_I(-\omega) = -X_I(\omega)$  (odd)

$$|X(-\omega)| = |X(\omega)| \text{ (even)}$$
  
$$\angle X(-\omega) = -\angle X(\omega) \text{ (odd)}$$

Real and even signals.  $x_R(n) = x(n)$ ,  $x_I(n) = 0$  and x(-n) = x(n).

$$\begin{array}{lll} X_R(\omega) &=& x(0)+2\sum_{n=1}^\infty x(n)\cos\omega n \quad \mbox{(even)} \\ X_I(\omega) &=& 0 \end{array}$$

It has real-valued spectrum, which is even function of the frequency  $\omega$ .

### Symmetry Properties of the Fourier Transform

Real and odd signals.  $x_R(n) = x(n)$ ,  $x_I(n) = 0$  and x(-n) = -x(n).

$$X_R(\omega) = 0$$
  

$$X_I(\omega) = -2\sum_{n=1}^{\infty} x(n) \sin \omega n \quad \text{(odd)}$$

It has imaginary-valued spectrum, which is odd function of the frequency  $\omega$ .

Purely imaginary signals.  $x_R(n) = 0$  and  $jx_I(n) = x(n)$ .

$$X_R(\omega) = \sum_{n=-\infty}^{\infty} x_I(n) \sin \omega n \quad \text{(odd)}$$
$$X_I(\omega) = \sum_{n=-\infty}^{\infty} x_I(n) \cos \omega n \quad \text{(even)}$$

### Symmetry Properties of the Fourier Transform

Purely imaginary and odd signals.  $x_R(n) = 0$ ,  $jx_I(n) = x(n)$  and  $x_I(-n) = -x_I(n)$ .

$$X_R(\omega) = 2\sum_{n=1}^{\infty} x_I(n) \sin \omega n \quad (\text{odd})$$
$$X_I(\omega) = 0$$

It has real-valued spectrum, which is odd function of the frequency  $\boldsymbol{\omega}.$ 

Purely imaginary and even signals.  $x_R(n) = 0$ ,  $jx_I(n) = x(n)$  and  $x_I(-n) = x_I(n)$ .

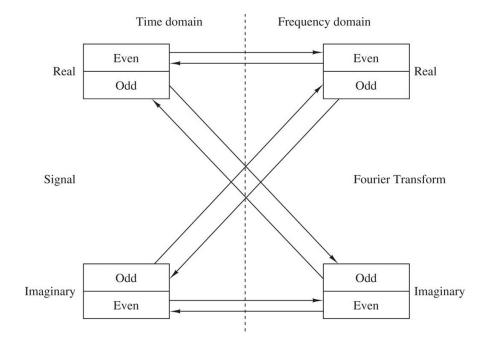
$$X_R(\omega) = 0$$
  

$$X_I(\omega) = x_I(0) + 2\sum_{n=1}^{\infty} x_I(n) \cos \omega n \quad (even)$$

It has imaginary-valued spectrum, which is even function of the frequency  $\omega$ .

### Symmetry Properties of the Fourier Transform

#### Summary of symmetry properties for the Fourier Transform



Linearity.

If  $x_1(n) \longleftrightarrow X_1(\omega)$  and  $x_2(n) \longleftrightarrow X_2(\omega)$ ,

then

$$\alpha_1 x_1(n) + \alpha_2 x_2(n) \longleftrightarrow \alpha_1 X_1(\omega) + \alpha_2 X_2(\omega).$$

## Fourier Transform Theorems and Properties

Time shifting.

If 
$$x(n) \longleftrightarrow X(\omega)$$
,

then

 $x(n-k) \longleftrightarrow e^{-j\omega k} X(\omega).$ 

Time reversal.

If 
$$x(n) \longleftrightarrow X(\omega)$$
,

then

$$x(-n) \longleftrightarrow X(-\omega).$$

## Fourier Transform Theorems and Properties

Convolution theorem.

If 
$$x_1(n) \longleftrightarrow X_1(\omega)$$
 and  $x_2(n) \longleftrightarrow X_2(\omega)$ ,

$$x(n) = x_1(n) \otimes x_2(n) \longleftrightarrow X(\omega) = X_1(\omega)X_2(\omega).$$

Correlation theorem.

If 
$$x_1(n) \longleftrightarrow X_1(\omega)$$
 and  $x_2(n) \longleftrightarrow X_2(\omega)$ ,

then

 $r_{x_1x_2}(l) \longleftrightarrow S_{x_1x_2}(\omega) = X_1(\omega)X_2(-\omega).$ 

### Fourier Transform Theorems and Properties

The Wiener-Khintchine theorem.

If x(n) is a real signal, then  $r_{xx}(l) \longleftrightarrow S_{xx}(\omega)$ .

Notice that neither the autocorrelation nor the energy spectral density has any phase information.

Frequency shifting.

If  $x(n) \longleftrightarrow X(\omega)$ ,

then

 $e^{j\omega_0 n} x(n) \longleftrightarrow X(\omega - \omega_0).$ 

## Fourier Transform Theorems and Properties

The modulation theorem.

If 
$$x(n) \longleftrightarrow X(\omega)$$
,

$$x(n)\cos\omega_0 n \longleftrightarrow \frac{1}{2}[X(\omega+\omega_0)+X(\omega-\omega_0)].$$

Parseval's theorem.

If 
$$x_1(n) \longleftrightarrow X_1(\omega)$$
 and  $x_2(n) \longleftrightarrow X_2(\omega)$ ,

then

$$\sum_{n=-\infty}^{\infty} x_1(n) x_2^*(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} X_1(\omega) X_2^*(\omega) d\omega.$$

## Fourier Transform Theorems and Properties

Windowing theorem.

If 
$$x_1(n) \longleftrightarrow X_1(\omega)$$
 and  $x_2(n) \longleftrightarrow X_2(\omega)$ ,

$$x_1(n)x_2(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} X_1(\lambda)X_2(\omega - \lambda)d\lambda.$$

Differentiation in the frequency domain.

If  $x(n) \longleftrightarrow X(\omega)$ ,

$$nx(n) \longleftrightarrow j \frac{dX(\omega)}{d\omega}.$$